Sequential network examples in VHDL, hardware implementation of dataflow models

Tutorial 05 on Dedicated systems

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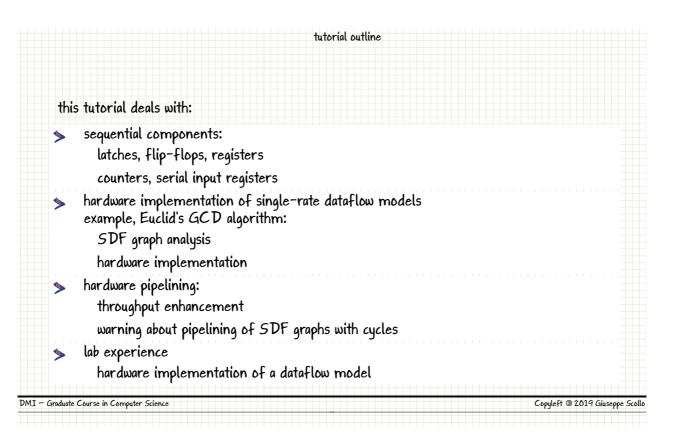
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1. Sequential network examples in VHDL, hardware implementation of dataflow models

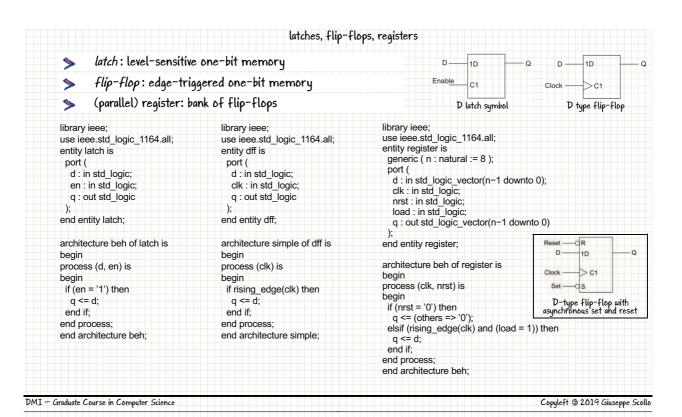
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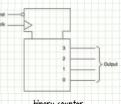


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counters, serial input registers

- counters: registers counting specified clock edges also used to implement timers
- serial input registers: partly similar to counters used for data input from serial lines, output is parallel the use of variables easies the VHDL description in behavioural style



```
use ieee.std_logic_1164.all;
use ieee.numeric_std.all;
entity counter is
 generic ( n : integer := 4 );
 port (
clk : in std_logic;
  rst : in std_logic
  output : out std_logic_vector(n-1 downto 0)
end;
architecture simple of counter is
begin
 process(clk, rst)
  variable count : unsigned(n-1 downto 0);
 begin
  if rst = '0' then
  count := (others => '0');
elsif rising_edge(clk) then
count := count + 1;
  output <= std_logic_vector(count);
 end process;
```

```
binary counter
use ieee.std_logic_1164.all;
entity shift_register is
generic ( n : integer := 4 );
 port (
clk : in std_logic;
rst : in std_logic;
   din : in std_logic;
   q : out std_logic_vector(n-1 downto 0)
end entity;
architecture simple of shift_register is
begin
 process(clk, rst)
   variable shift_reg : std_logic_vector(n-1 downto 0);
  begin
   if rst = '0' then
   shift_reg := (others => '0');
elsif rising_edge(clk) then
shift_reg := shift_reg(n-2 downto 0) & din;
   q <= shift_reg;
 end process;
end architecture simple;
```

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single-rate SDF graph to hardware

hardware implementation assumption:

single-rate SDF graphs, all actors operate at the same clock frequency

three implementation rules:

- 1. all actors are implemented as combinational circuits
- all communication queues are implemented as wires (without storage)
- 3. each initial token on a communication queue is replaced by a register

two definitions:

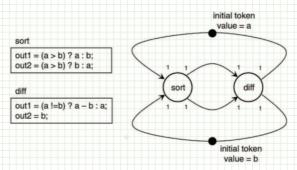
- > combinational path in the SDF graph: cycle-free path with no initial tokens on it
- critical path in the SDF graph: combinational path that has maximum latency

maximum clock frequency for the circuit: reciprocal of latency through critical path

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example: Euclid's GCD algorithm, SDF graph analysis

algorithm: at each step (a, b) is replaced by (|a-b|, min(a,b)) the pair converges to (GCD(a,b), GCD(a,b))



Schaumont, Figure 3.10 - Euclid's greatest common divisor as an SDF graph

PASS analysis:

$$G = \begin{bmatrix} +1 & -1 \\ +1 & -1 \\ -1 & +1 \\ -1 & +1 \\ -1 & +1 \end{bmatrix} \leftarrow edge(sort, diff) \\ \leftarrow edge(sort, diff) \\ \leftarrow edge(diff, sort)$$
 rank(G) = 1 $q_{PASS} = \begin{bmatrix} 1 \\ 1 \end{bmatrix}$

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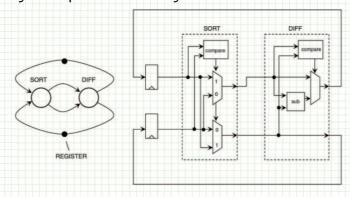
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hardware implementation of Euclid's GCD algorithm

by the aforementioned three rules for a hardware implementation of the SDF model:

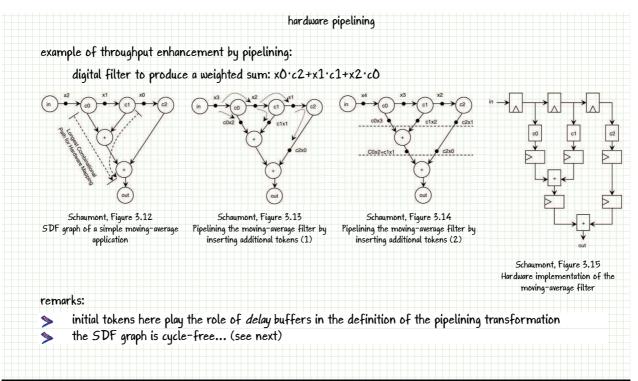
- two combinational circuits implement the actors
- > a register is placed on each of the two connections from diff to sort implementing the actors is a simple matter, by means of a few commonly used modules (multiplexers, comparators and a subtractor)

N.B. the HW diagram requires a bit of imagination and a correction



Schaumont, Figure 3.11 - Hardware implementation of Euclid's algorithm

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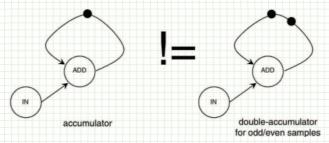
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pipelining in SDF graphs with loops

by introducing new tokens, pipelining may change the behaviour of an SDF graph in particular, this may happen if additional tokens are introduced inside a loop, as this example shows:



Schaumont, Figure 3.16 - Loops in SDF graphs cannot be pipelined

in order to apply pipelining without changing the functional behaviour of an SDF graph with cycles, the additional tokens should be placed outside of any loop in the graph for instance, on the input or output lines

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lab experience

the circuit depicted in figure 3.11 implements the computational core of Euclid's GCD algorithm, yet it does not contain elements apt to signal the start and the end of the computation nor to distinguish inputs and output; the aims of this experience are: to extend that circuit to this purpose, to produce a VHDL description of it, to simulate it, and to implement it on the DE1-SoC FPGA

1. extend the schematic of the circuit in figure 3.11 with three input signals and two output signals:

a, b: the input data, 5-bit wide each

start: 1-bit input, to signal availability of the input data

gcd: the 5-bit output result

done: 1-bit output, to signal the end of computation and availability of the output result

and with additional elements (flip-flop, multiplexers, maybe a comparator) useful to the stated purpose

- produce a NHDL description of the designed circuit, either directly or through a Gezel description to be translated to NHDL by means of the fdlvhd program
- 3. create a Quartus project Euclid, assign it the produced .vhd files, compile, and simulate the behaviour of the circuit with a few input data pairs
- 4. create a new Quartus project Euclid_on_DE1SoC, assign it the previous .vhd files together with the 7-segment display decoder employed in lab tutorial 4 and a new top-level NHDL entity, composing the former one with an instance of the aforementioned decoder, while mapping the I/O signals to FPGA pins as follows:

a, b: SW9-5, SW4-0

start: not KEY1

RST: not KEY0

CLK: CLOCK_50 (50 MHz system clock)

done: LEDR0 gcd: LEDR1, HEX0

5. import the DE1-SoC pin assignments, compile, program the FPGA with the resulting .sof file, and test the functioning of the implementation with a few input data pairs

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references

recommended readings:

Zwoliński, Ch. 6, Sect. 6.1-6.5.1

Schaumont, Ch. 3, Sect 3.2

readings for further consultation:

Schaumont, Ch. 3, Sect. 3.3

useful materials for the proposed lab experience

(source: Intel Corp. - FPGA University Program, 2016)

Debugging of VHDL Hardware Designs on Intel's DE-Series Boards - For Quartus Prime 16.1, Sect. 4.1, 6-8

VHDL sources:

examples in this presentation examples in Zwoliński's book

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